Microwave Engineering

5-2

Chapter 5 Impedance Matching

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Learning Objectives

- Understanding about single stub matching
- Learn how to match impedance using Smith chart
- Learn how to match impedance using mathematical equations
- Practice impedance matching with examples

Learning contents

- Introduction about Single Stub Matching
- Shunt- and Series-Stub Matching using Graphical Solution
- Shunt- and Series-Stub Matching using Mathematical Solution
- Impedance Matching Example using Calculation and Smith Chart

1 Introduction about Single Stub Matching

• Single stub: single open- or short-circuited length of transmission line (a 'stub'), connected either in parallel or in series with the transmission line at a certain distance from load.



- Shunt stub matching is easy relatively than series stub in realization of microstrip line or strip line forms.
- Open-circuited stubs are easier to fabricate since a via hole through the substrate to ground plane is not needed
- Shunt stub case matching
 - $\begin{array}{ccc} & Y_L \longrightarrow Y_0 + jB \\ & \longrightarrow & Y_0 \end{array}$
- (by using transmission line with length *d*) (by using transmission line having length *l*)
- Series stub case matching

$$\begin{array}{c} - Z_L \to Z_0 + jX \\ \to Z_0 \end{array}$$

 Depending on the length and direction of transmission line, load reflection coefficient rotates along WTG or WTL with same radius on Smith chart.
 Wavelengths toward generator







- Mathematical evaluation of shunt-stub matching
 - Derivation of formulas for d and l
 - Load impedance (or admittance)

$$Z_L = 1/Y_L = R_L + jX_L$$

- Impedance Z down a length, d, of transmission line from load



$$Z_0 R_L (1+t^2) = R_L^2 + (X_L + Z_0 t)^2$$

$$Z_0 (R_L - Z_0) t^2 - 2X_L Z_0 t + (R_L Z_0 - R_L^2 - X_L^2) = 0$$

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- (continued)

$$Z_{0}(R_{L} - Z_{0})t^{2} - 2X_{L}Z_{0}t + (R_{L}Z_{0} - R_{L}^{2} - X_{L}^{2}) = 0$$

$$t = \tan \beta d = \tan \frac{2\pi}{\lambda} d = \frac{X_{L} \pm \sqrt{R_{L}[(Z_{0} - R_{L})^{2} + X_{L}^{2}]/Z_{0}}}{R_{L} - Z_{0}} = f(Z_{0}, R_{L}, X_{L}) : \text{known values} \quad @R_{L} \neq Z_{0}$$

$$\frac{d}{\lambda} = \begin{cases} \frac{1}{2\pi} \tan^{-1} t, & \text{for } t \ge 0 \\ \frac{1}{2\pi} (\pi + \tan^{-1} t), & \text{for } t < 0. \end{cases}$$

$$\text{If } R_{L} = Z_{0}, \text{ then } t = -X_{L} / 2Z_{0}.$$

$$\text{To find required stub lengths}, B_{S} = -B.$$

1) For an open-circuited stub,

$$jB_{S} = jY_{0} \tan\beta l_{o} = jY_{0} \tan(2\pi l_{o}/\lambda) = -jB$$
$$\frac{l_{o}}{\lambda} = \frac{1}{2\pi} \tan^{-1} \left(\frac{B_{S}}{Y_{0}}\right) = \frac{-1}{2\pi} \tan^{-1} \left(\frac{B}{Y_{0}}\right)$$

2) For a short-circuited stub,

$$jB_{S} = -jY_{0}\cot\beta l_{s} = -jY_{0}\cot(2\pi l_{s}/\lambda) = -jB$$
$$\frac{l_{s}}{\lambda} = \frac{-1}{2\pi}\tan^{-1}\left(\frac{Y_{0}}{B_{s}}\right) = \frac{1}{2\pi}\tan^{-1}\left(\frac{Y_{0}}{B}\right)$$

 \rightarrow If the lengths (l_o, l_s) are negative, $\lambda/2$ can be added to give positive results.

- Mathematical evaluation of series-stub matching
 - Derivation of formulas for d and l
 - Load admittance (or impedance)

$$Y_L = 1/Z_L = G_L + jB_L$$

- Admittance Y down a length, d, of transmission line from load

$$Y = Y_0 \frac{Y_L + jY_0 \tan \beta d}{Y_0 + jY_L \tan \beta d} = Y_0 \frac{(G_L + jB_L) + jY_0 t}{Y_0 + j(G_L + jB_L)t}$$

where $t = \tan \beta d$, $Y_0 = 1/Z_0$
$$Z = \frac{1}{Y} = R + jX = Z_0 + jX$$

where $R = \frac{G_L(1 + t^2)}{G_L^2 + (B_L + Y_0 t)^2} = Z_0 = \frac{1}{Y_0}$ and $X = \frac{G_L^2 t - (Y_0 - tB_L)(B_L + tY_0)}{Y_0[G_L^2 + (B_L + Y_0 t)^2]}$



$$Y_0 G_L (1 + t^2) = G_L^2 + (B_L + Y_0 t)^2$$

$$Y_0 (G_L - Y_0) t^2 - 2B_L Y_0 t + (G_L Y_0 - G_L^2 - B_L^2) = 0$$

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- (continued)

$$Y_{0}(G_{L} - Y_{0})t^{2} - 2B_{L}Y_{0}t + (G_{L}Y_{0} - G_{L}^{2} - B_{L}^{2}) = 0$$

$$t = \tan \beta d = \tan \frac{2\pi}{\lambda} d = \frac{B_{L} \pm \sqrt{G_{L}[(Y_{0} - G_{L})^{2} + B_{L}^{2}]/Y_{0}}}{G_{L} - Y_{0}} = f(Y_{0}, G_{L}, B_{L}) : \text{known values} \quad @G_{L} \neq Y_{0}$$

$$\frac{d}{\lambda} = \begin{cases} \frac{1}{2\pi} \tan^{-1} t, & \text{for } t \ge 0 \\ \frac{1}{2\pi} (\pi + \tan^{-1} t), & \text{for } t < 0. \end{cases}$$

- If $G_L = Y_0$, then $t = -B_L / 2Y_0$.
- To find the required stub lengths, $X_S = -X$.
- 1) For a short-circuited stub,

$$jX_{S} = jZ_{0} \tan\beta l_{s} = jZ_{0} \tan(2\pi l_{s}/\lambda) = -jX$$
$$\frac{l_{s}}{\lambda} = \frac{1}{2\pi} \tan^{-1}\left(\frac{X_{S}}{Z_{0}}\right) = \frac{-1}{2\pi} \tan^{-1}\left(\frac{X}{Z_{0}}\right)$$



2) For an open-circuited stub, $jX_S = -jZ_0 \cot\beta l_o = -jZ_0 \cot(2\pi l_o/\lambda) = -jX$ $\frac{l_o}{\lambda} = \frac{-1}{2\pi} \tan^{-1} \left(\frac{Z_0}{X_s} \right) = \frac{1}{2\pi} \tan^{-1} \left(\frac{Z_0}{X} \right)$

 \rightarrow If the lengths (l_o, l_s) are negative, $\lambda/2$ can be added to give positive results.

• Ex. 1] Shunt single stub matching example

Design two single-stub (short circuit) shunt tuning networks to match a load impedance $Z_L = 80 + j100 [\Omega]$ to 50 [Ω] line at a frequency of 3 GHz. Plot the return loss in dB from 2 to 4 GHz for each solution.

1) Theoretical solutions

$$t = \frac{X_L \pm \sqrt{R_L[(Z_0 - R_L)^2 + X_L^2]/Z_0}}{R_L - Z_0} = \frac{100 \pm \sqrt{80[(50 - 80)^2 + 100^2]/50}}{80 - 50} = \begin{cases} 7.74 \text{ for "+" sign} \\ -1.07 \text{ for "-" sign} \end{cases}$$

$$B = \frac{R_L^2 t - (Z_0 - X_L t)(X_L + Z_0 t)}{Z_0[R_L^2 + (X_L + Z_0 t)^2]} = \begin{cases} 0.033 \text{ for } t = 7.74 \\ -0.033 \text{ for } t = -1.07 \end{cases}$$

$$d = \begin{cases} [\frac{1}{2\pi} \tan^{-1}(t)]\lambda = \frac{1}{2\pi} \tan^{-1}(7.74)\lambda = 0.23\lambda = 82.8^{\circ} \qquad (a \ t = 7.74 \\ \frac{1}{2\pi} [\pi + \tan^{-1}(t)]\lambda = \frac{1}{2\pi} [\pi + \tan^{-1}(-1.07)]\lambda = 0.37\lambda = 133.2^{\circ} \qquad (a \ t = -1.07 \\ 0.0867\lambda = 31.21^{\circ} \qquad (a \ B = 0.033 \\ -0.0867\lambda (a \ B = -0.033 (negative, by adding \lambda/2 \\ -\lambda l_s = 0.413\lambda = 148.68^{\circ}) \end{cases}$$

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2) Smith chart solutions

- $Z_L = 80 + j100$
- Normalized impedance: $z_L = Z_L/Z_0 = 1.6 + j2$
- Mark z_L and draw a reflection coefficient circle
- Convert to z_L to load admittance, y_L
- $|\Gamma|$ circle intersects the $1 \pm jb$ circle at two points denoted as y_1 and $y_2 \rightarrow d_1$ or d_2

$$d_1 = [(0.5 - 0.45) + 0.18]\lambda = 0.23\lambda \rightarrow 0.23 \times 360^\circ = 82.8^\circ$$

$$d_2 = [(0.5 - 0.45) + 0.32]\lambda = 0.37\lambda \rightarrow 0.37 \times 360^{\circ} = 133.2^{\circ}$$

- At two intersections point,

 $y_1 = 1 + j1.65, \quad y_2 = 1 - j1.65$

- By moving from $y = \infty$ (short-circuited) toward generator,

 $l_1 = (0.3367 - 0.25)\lambda = 0.0867\lambda \rightarrow 31.21^\circ, \quad l_2 = [(0.5 - 0.25) + 0.163]\lambda = 0.413\lambda \rightarrow 148.68^\circ$

➔ The matching stub is kept as short as possible to improve the matching bandwidth and reduce losses caused by two lossy transmission lines.





- Solution #01

• Ex. 2] Series single stub matching example

Design a single series open-circuited stub network to match a load impedance $Z_L = 75 + j100 [\Omega]$ to 50 [Ω] line at a frequency of 3 GHz. Plot the return loss in dB from 2 to 4 GHz for each solution.

1) Theoretical solutions

- Load admittance

$$Y_{L} = G_{L} + jB_{L} = \frac{1}{Z_{L}} = \frac{1}{75 + j100} = 0.0048 - j0.0064$$

$$- Y_{0} = \frac{1}{Z_{0}} = 0.02$$

$$- t = \frac{B_{L} \pm \sqrt{G_{L}[(Y_{0} - G_{L})^{2} + B_{L}^{2}]/Y_{0}}}{G_{L} - Y_{0}} = \frac{-0.0064 \pm \sqrt{0.0048[(0.02 - 0.0048)^{2} + (-0.0064)^{2}]/0.02}}{0.0048 - 0.02} = \begin{cases} -0.111 \text{ for "+" sign}}{0.953 \text{ for "-" sign}} = \begin{cases} -0.111 \text{ for "+" sign}}{0.953 \text{ for "-" sign}} = \begin{cases} -0.111 \text{ for "+" sign}}{0.953 \text{ for "-" sign}} = \begin{cases} -0.111 \text{ for "+" sign}}{0.953 \text{ for "-" sign}} = \begin{cases} -0.111 \text{ for "+" sign}}{0.953 \text{ for "-" sign}} = \begin{cases} -0.111 \text{ for "+" sign}}{0.953 \text{ for "-" sign}} = \begin{cases} -0.111 \text{ for "-" sign}}{0.953 \text{ for "-" sign}} = \begin{cases} -0.111 \text{ for "-" sign}}{0.953 \text{ for "-" sign}} = \begin{cases} -0.111 \text{ for "-" sign}}{0.953 \text{ for "-" sign}} = \begin{cases} -0.111 \text{ for "-" sign}}{0.953 \text{ for "-" sign}} = \begin{cases} -0.111 \text{ for "-" sign}}{0.953 \text{ for "-" sign}} = \begin{cases} -0.111 \text{ for "-" sign}}{0.953 \text{ for "-" sign}} = \begin{cases} -0.111 \text{ for "-" sign}}{0.953 \text{ for "-" sign}} = \begin{cases} -0.111 \text{ for "-" sign}}{0.953 \text{ for "-" sign}} = \begin{cases} -0.111 \text{ for "-" sign}}{0.953 \text{ for "-" sign}} = \begin{cases} -0.111 \text{ for "-" sign}}{0.953 \text{ for "-" sign}} = \begin{cases} -0.111 \text{ for "-" sign}}{0.953 \text{ for "-" sign}} = \begin{cases} -0.111 \text{ for "-" sign}}{0.953 \text{ for "-" sign}} = \begin{cases} -0.111 \text{ for "-" sign}}{0.953 \text{ for "-" sign}} = \begin{cases} -0.111 \text{ for "-" sign}}{0.953 \text{ for "-" sign}} = \begin{cases} -0.111 \text{ for "-" sign}}{0.953 \text{ for "-" sign}} = \begin{cases} -0.111 \text{ for "-" sign}} = \begin{cases} -0.111 \text{ for "-" sign}}{0.953 \text{ for "-" sign}} = \begin{cases} -0.111 \text{ for "-" sign}} = \end{cases}$$

- Length from load to stub:

$$d = \begin{cases} \frac{1}{2\pi} [\pi + \tan^{-1}(t)] \lambda = \frac{1}{2\pi} [\pi + \tan^{-1}(-0.111)] \lambda = 0.482\lambda = 173.52^{\circ} \quad @ t = -0.111 \\ [\frac{1}{2\pi} \tan^{-1}(t)] \lambda = \frac{1}{2\pi} \tan^{-1}(0.953) \lambda = 0.121\lambda = 43.56^{\circ} \quad @ t = 0.935 \end{cases}$$

- For short-circuited stub

$$l_{s} = \frac{1}{2\pi} \tan^{-1} \left(\frac{Y_{0}}{B} \right) \lambda = \begin{cases} 0.085\lambda \to 30.6^{\circ} & a X = 84.10 \\ -0.085\lambda & a X = -84.10 \text{ (negative, by adding } \lambda / 2 \to l_{s} = 0.415\lambda \to 149.4^{\circ} \text{)} \end{cases}$$

- Therefore,

 $d_1 = 0.121\lambda \rightarrow 43.56^\circ, \ l_{s1} = 0.415\lambda \rightarrow 149.4^\circ$ $d_2 = 0.482\lambda \rightarrow 173.52^\circ, \ l_{s2} = 0.085\lambda \rightarrow 30.6^\circ$

2) Smith chart solutions

- $Z_L = 75 + j100$
- Normalized impedance: $z_L = Z_L/Z_0 = 1.5 + j2$
- Mark z_L and draw a reflection coefficient circle
- $|\Gamma|$ circle intersects the 1 + jx circle at two points denoted as z_1 and $z_2 \rightarrow d_1$ or d_2

$$d_1 = (0.319 - 0.198)\lambda = 0.121\lambda \rightarrow 43.56^{\circ}$$

$$d_2 = [(0.5 - 0.198) + 0.18)]\lambda = 0.482\lambda \rightarrow 173.52^{\circ}$$

- At the two intersections point,

 $z_1 = 1 - j1.69, \quad z_2 = 1 + j1.69$

- By moving from $z = \infty$ (open-circuited) toward generator

$$l_1 = [(0.5 - 0.25)] + 0.165\lambda = 0.415\lambda \rightarrow 149.4^{\circ}$$
$$l_2 = (0.335 - 0.25)\lambda = 0.085\lambda \rightarrow 30.6^{\circ}$$







5 Review

- Single stub: single open- or short-circuited length of transmission line (a 'stub'), connected either in parallel or in series with the transmission line at a certain distance from the load.
- Shunt stub is easier to realization than series stub and open stub is easier to fabrication than short stub.



Admittance (or Impedance) Smith Chart