**Microwave Engineering 6-1** 



# **Chapter 6 Microwave Resonators**

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#### **Learning Objectives**

- Understanding overview of resonators
- Understanding parallel LC resonant circuit
- Understanding series LC resonant circuit

#### **Learning contents**

- Introduction to Microwave Resonators
- Parallel LC Resonant Circuit
- Series LC Resonant Circuit

#### **1 Introduction To Microwave Resonators**

- Microwave resonators are used in a variety of applications such as:
	- Filters
	- Oscillators
	- Frequency meters
	- Tuned amplifiers etc.
	-
- The operation of microwave resonators has a similar circuit theory to that of the lumped-element resonators, thus the basic of series and parallel RLC resonant circuits will be reviewed firstly and shortly.
- Types of resonator:
	- Lumped resonator or RLC resonator  $\rightarrow$  Used for low frequency<br>- Transmission line resonator  $\rightarrow$  Used for microwave frequency
	-
	- Cavity resonator → Used for RF frequency and high power handling application dye to high quality factor.

- § As its name, a parallel *LC* resonant circuit (or *LC*-tank) is an electrical circuit consisting of an inductor (*L*) and a capacitor (*C*) connected in parallel, which maintain the same voltage across the components.
	- Input impedance

$$
Z_{\text{in}} = \left(\frac{1}{R} + \frac{1}{j\omega L} + j\omega C\right)^{-1}
$$

- Complex power delivered to resonator

$$
P_{\text{in}} = \frac{1}{2}VI^* = \frac{1}{2}Z_{\text{in}} |I|^2 = \frac{1}{2}|V|^2 \frac{1}{Z_{\text{in}}^*} = \frac{1}{2}|V|^2 Y_{\text{in}}^*
$$
\n
$$
= \frac{1}{2}|V|^2 \left(\frac{1}{R} - \frac{1}{j\omega L} - j\omega C\right)
$$
\n
$$
= \frac{1}{2} \frac{|V|^2}{R} + j2\omega \left(\frac{1}{4} \frac{|V|^2}{\omega^2 L} - \frac{1}{4}|V|^2 C\right)
$$
\n
$$
= P_l + j2\omega (W_m - W_e)
$$



$$
P_{l} = \frac{1}{2} \frac{|V|^{2}}{R} = \frac{1}{2} G |V|^{2}
$$

$$
W_e = \frac{1}{4} |V|^2 C = \frac{1}{4} V V^* C
$$

$$
W_m = \frac{1}{4} |I_L|^2 L = \frac{1}{4} I_L I_L^* L = \frac{1}{4} \left| \frac{V}{\omega L} \right|^2 L = \frac{1}{4 \omega^2 L} V V^*
$$

EXECUTE: The expression is represented by resistor *R*.

\n
$$
P_{i} = \frac{1}{2} \frac{|V|^{2}}{R} = \frac{1}{2} G |V|^{2}
$$
\naverage electric energy stored in capacitor *C*

\n
$$
W_{e} = \frac{1}{4} |V|^{2} C = \frac{1}{4} V V^{*} C
$$
\naverage magnetic energy stored in inductor *L*

\n
$$
W_{m} = \frac{1}{4} |I_{L}|^{2} L = \frac{1}{4} I_{L} I_{L}^{*} L = \frac{1}{4} \frac{|V|^{2}}{\omega L} L = \frac{1}{4 \omega^{2} L} V V^{*}
$$
\nput impedance

\n
$$
P_{in} = \frac{1}{2} V I^{*} = \frac{1}{2} Z_{in} |I|^{2}
$$
\n⇒  $Z_{in} = \frac{P_{in}}{1} = \frac{P_{i} + j2\omega (W_{m} - W_{e})}{\frac{1}{2} |I|^{2}}$ 



**Parallel LC Resonant Circuit**  
\n- If 
$$
W_e = W_m
$$
  
\n
$$
\frac{1}{4} V V^* C = \frac{1}{4\omega_0^2 L} V V^* \rightarrow C = \frac{1}{\omega_0^2 L}
$$
\n
$$
\therefore \ \omega = \omega_0 = \sqrt{\frac{1}{LC}}
$$
\nQuality factor (Q-factor)

- $=\omega_0 \frac{\text{average stored energy}}{112.5}$  $Q = \omega_0 \frac{\text{average stored energy}}{\text{dissipated power loss}}$
- 

2 **Parallel LC Resonant Circuit**  
\n- If 
$$
W_e = W_m
$$
  
\n
$$
\frac{1}{4}VV^*C = \frac{1}{4\omega_0^2 L}VV^* \rightarrow C = \frac{1}{\omega_0^2 L}
$$
\n
$$
\therefore \omega = \omega_0 = \sqrt{\frac{1}{LC}}
$$
\n2 **Quality factor (Q-factor)**  
\n2 **Quality factor (Q-factor)**  
\n2 **Required power loss**  
\n $Q = \omega_0$   $\frac{\text{average stored energy}}{\text{disipated power loss}}$   
\n- Stored time average energy  
\n $W = W_m + W_e = 2W_m = 2W_e = \frac{1}{2}CVV^*$   
\n- Disipated power:  $P_i = \frac{1}{2}GVV^*$   
\n $\Rightarrow Q = \omega_0 \frac{\frac{1}{2}CVV^*}{\frac{1}{2}GVV^*} = \frac{\omega_0 C}{G} = \omega_0 RC = \frac{R}{\omega_0 L}$  (in parallel RLC network)

![](_page_5_Figure_5.jpeg)

■ Impedance at around resonance frequency (
$$
\omega
$$
  $\omega = \omega_0 \pm \Delta \omega$ )

**Parallel LC Resonant Circuit**  
\nImpedance at around resonance frequency (
$$
\omega \omega = \omega_0 \pm \Delta \omega
$$
)  
\n
$$
Z_{in} = \left[ \frac{1}{R} + \frac{1}{j\omega L} + j\omega C \right]^{-1} = \frac{1}{\frac{1}{R} + \frac{1}{j\omega L} + j\omega C}
$$
\n
$$
= \frac{R}{1 + j\omega_0 RC \left( \frac{\omega}{\omega_0} - \frac{1}{\omega_0 \omega LC} \right)} = \frac{R}{1 + jQ \left( \frac{\omega}{\omega_0} - \frac{\omega_0}{\omega} \right)}
$$
\n
$$
= \frac{R}{1 + jQ \frac{(\omega + \omega_0)(\omega - \omega_0)}{\omega \omega_0}} \approx \frac{R}{1 \pm j2Q \frac{\Delta \omega}{\omega_0}}
$$
\n
$$
= |Z_{in}|_{max} = R \quad \text{and} \quad \omega = \omega_0
$$
\n
$$
= \frac{1}{1 + 2Q \frac{\Delta \omega}{\omega_0}} = 1, \quad Q = \frac{\omega_0}{2\Delta \omega} \text{ and } |Z_{in}| = \frac{1}{\sqrt{2}} |Z_{in}|_{max}
$$
\nwhere 2 $\Delta \omega$ : bandwidth, 2 $\Delta \omega / \omega_0$ : fractional bandwidth (FBW)

- If 
$$
2Q \frac{\Delta \omega}{\omega_0} = 1
$$
,  $Q = \frac{\omega_0}{2\Delta \omega}$  and  $|Z_{in}| = \frac{1}{\sqrt{2}} |Z_{in}|_{max}$ 

![](_page_6_Figure_5.jpeg)

$$
Q = \frac{\omega_0}{2\Delta\omega} = \frac{1}{\text{FBW}}
$$

-  $Q$  including load or source resistance:  $Q_L$  (loaded  $Q$ )

$$
Q = \frac{\omega_0}{2\Delta\omega} = \frac{1}{FBW}
$$
  
\n- Q including load or source resistance:  $Q_L$  (loaded Q)  
\n
$$
Q_L = \frac{R/|R_L|}{\omega_0 L} = \frac{RR_L/(R+R_L)}{\omega_0 L}
$$
  
\n- If  $R = \infty$ , the external Q can be calculated as  
\n
$$
Q_{ex} = \frac{R_L}{\omega_0 L}
$$
  
\n- If  $R_L = \infty$ , the internal Q or unloaded Q is

- If  $R = \infty$ , the external Q can be calculated as

$$
Q_{ex} = \frac{R_L}{\omega_0 L}
$$

$$
Q_{\rm in} = Q_u = \frac{R}{\omega_0 L}
$$

![](_page_7_Figure_9.jpeg)

■ General relationship among Q-factors

**arallel LC Resonant Circuit**  
\n
$$
Q_{\lambda} = \frac{R/R_{\lambda}}{\omega_{0}L} = \frac{RR_{\lambda}/(R + R_{\lambda})}{\omega_{0}L}
$$
\n
$$
Q_{\lambda} = \frac{R/R_{\lambda}}{R R_{\lambda}} = \omega_{0}L \frac{R + R_{\lambda}}{R R_{\lambda}} = \omega_{0}L(\frac{1}{R} + \frac{1}{R_{\lambda}})
$$
\n
$$
= \frac{\omega_{0}L}{R} + \frac{\omega_{0}L}{R_{\lambda}} = \frac{1}{Q_{\mu}} + \frac{1}{Q_{\mu}} \Rightarrow Q_{\lambda} < Q_{\mu} \& Q_{\mu}
$$
\n**energy damping factor** ( $\delta$ ): Decaying ratio of oscillation amplitude in case of drop of external energy\n
$$
W = W_{0}e^{-2\delta t}
$$
\n
$$
- \frac{dW}{dt} = P_{\lambda} = 2\delta W
$$
\n
$$
\delta = \frac{P_{\lambda}}{2W} = \frac{\omega_{0}}{2} \frac{P_{\lambda}}{\omega_{0}W} = \frac{\omega_{0}}{2Q}
$$
\n
$$
W = W_{0}e^{-\omega_{0}/Q} \rightarrow Higher Q - factor guarantees longer energy conservation!
$$

![](_page_8_Figure_3.jpeg)

**Energy damping factor** ( $\delta$ ) : Decaying ratio of oscillation amplitude in case of drop of external energy  $2\delta t$  $0<sup>c</sup>$ C Resonant Circuit<br>
hip among Q-factors<br>  $\frac{R_k/(R+R_k)}{\omega_k L}$ <br>  $\frac{R_k R_k}{RR_k} = \omega_k L(\frac{1}{R} + \frac{1}{R_k})$ <br>  $\frac{1}{\frac{1}{Q_m} + \frac{1}{Q_m}} = Q_k < Q_m$  & Q<sub>av</sub><br>  $\frac{R_k}{R} = \frac{Q_k}{Q_m}$ <br>  $\frac{R_i}{\omega_k W} = \frac{\omega_b}{2Q}$ <br>  $\rightarrow Higher Q$  - factor guarantees longer e *t Higher Q factor gu*  $\delta t$  $-2\delta t$ Resonant Circuit<br>
among Q-factors<br>  $\frac{(R+R_L)}{RR_L} = \omega_0 L(\frac{1}{R} + \frac{1}{R_L})$ <br>  $\frac{1}{RR_L} + \frac{1}{Q_{\alpha}} \Rightarrow Q_L < Q_{\alpha} \& Q_{\alpha}$ <br>  $\frac{1}{Q_{\alpha}} + \frac{1}{Q_{\alpha}} \Rightarrow Q_L < Q_{\alpha} \& Q_{\alpha}$ <br>  $\cot \theta$  : Decaying ratio of oscillation amplitude in case of drop o *R R R*

$$
W = W_0 e^{-2\delta t}
$$
  
\n
$$
-\frac{dW}{dt} = P_l = 2\delta W
$$
  
\n
$$
\delta = \frac{P_l}{2W} = \frac{\omega_0}{2} \frac{P_l}{\omega_0 W} = \frac{\omega_0}{2Q}
$$
  
\n
$$
W = W_0 e^{-\omega_0 t/Q} \qquad \text{Kietov equations longer expression}
$$

 $v_0 t/Q \rightarrow Hichor$  $W_0e^{-\omega_0 t/Q}$   $\rightarrow$  *Higher Q* - factor guarantees longer energy convservation! ! The contract of the contract of the

#### **3 Series LC Resonant Circuit**

- <sup>■</sup> In series LC circuit, the components share the same current but have different voltage across each, showing voltage summation.
	- Input impedance

$$
Z_{\text{in}} = R + j\omega L - j\frac{1}{\omega C}
$$

- Complex power delivered to the resonator

**ries LC Resonant Circuit**  
\nseries LC circuit, the components share the same current but have different vo  
\ntage summation.  
\nbut impedance  
\n
$$
\frac{1}{\omega C}
$$
\nmplex power delivered to the resonator  
\n
$$
P_{\text{in}} = \frac{1}{2}VI^* = \frac{1}{2}Z_{\text{in}}|I|^2 = \frac{1}{2}|V|^2 \frac{1}{Z_{\text{in}}^*}
$$
\n
$$
= \frac{1}{2}|I|^2 \left(R + j\omega L - j\frac{1}{\omega C}\right)
$$
\n
$$
= \frac{1}{2}|I|^2 R + j2\omega \left(\frac{1}{4}|I|^2 L - \frac{1}{4} \frac{|I|^2}{\omega^2 C}\right)
$$
\n
$$
= P_l + j2\omega (W_m - W_e)
$$

![](_page_9_Figure_6.jpeg)

#### **Series LC Resonant Circuit**

$$
P_l = \frac{1}{2} |I|^2 R
$$

$$
W_e = \frac{1}{4} |V_C|^2 C = \frac{1}{4} |I|^2 \frac{1}{\omega^2 C}
$$

**eries LC Resonant Circuit**<br>
ver dissipated by resistor (R)<br>  $=\frac{1}{2}|I|^2 R$ <br>
rage electric energy stored in capacitor (C)<br>  $V_e = \frac{1}{4}|V_c|^2 C = \frac{1}{4}|I|^2 \frac{1}{\omega^2 C}$ <br>
where  $V_c$ : voltage across capacitor<br>
rage magnetic energy

$$
Z_{\text{in}} = \frac{P_{\text{in}}}{\frac{1}{2}|I|^2} = \frac{P_l + 2j\omega(W_m - W_e)}{\frac{1}{2}|I|^2}
$$

![](_page_10_Figure_9.jpeg)

- At the resonant frequency,  $W_m = W_e$ 

$$
\omega = \omega_0 = \frac{1}{\sqrt{LC}}
$$

- Quality factor (*Q*-factor) +
- Unloaded Q can be evaluated at resonance:  $W_m = W_e$

$$
Q_u = \omega_0 \frac{2W_m}{P_l} = \frac{\omega_0 L}{R} = \frac{1}{\omega_0 RC}
$$
  
\n
$$
\rightarrow Q \text{ is increased as } R \text{ decreases}
$$

- Impedance near the resonate frequency:  $\omega = \omega_0 \pm \Delta \omega$ 

$$
Z_{\text{in}} = R + j\omega_0 L \left(\frac{\omega}{\omega_0} - \frac{1}{\omega \omega_0 LC}\right) = R + j\omega_0 L \left(\frac{\omega^2 - \omega_0^2}{\omega_0 \omega}\right)
$$

![](_page_11_Figure_8.jpeg)

**Lumped Resonant Circuit: LC Series R**  
\n- Due to 
$$
\omega^2 - \omega_0^2 = (\omega - \omega_0)(\omega + \omega_0)
$$
  
\n $= \Delta \omega (2\omega - \Delta \omega)$   
\n $\approx 2\omega \Delta \omega$  for small  $\Delta \omega$ 

- Relation between  $Z_{\text{in}}$  and  $Q_u$ 

Lumped Resonant Circuit: LC Series Reson  
\n
$$
ω2 - ω02 = (ω - ω0)(ω + ω0)
$$
\n
$$
= Δω(2ω - Δω)
$$
\n
$$
= 2ωΔω \text{ for small } Δω
$$
\n
$$
Zin = R + jω0L \left( \frac{ω2 - ω02}{ω0ω} \right)
$$
\n
$$
= R \left\{ 1 + j \frac{ω0L}{R} \left( \frac{2ωΔω}{ω0ω} \right) \right\}
$$
\n
$$
= R \left\{ 1 + j2Q \frac{Δω}{ω0} \right\}
$$
\n(in series RLC network) (in series RLC network)

![](_page_12_Figure_4.jpeg)

13

- **Example:** A series resonant network consisting of a resistor of 30  $\Omega$ , a capacitor of 2  $\mu$ F, and an inductor of 2 mH is connected across a sinusoidal supply voltage which has a constant output of 9 V at all frequencies. **1 Resonant Circuit:** LC Series Resonant Circuit<br>series resonant network consisting of a resistor of 30  $\Omega$ , a capacitor of 2  $\mu$ F, and an inductor<br>neterd across a sinusoidal supply voltage which has a constant output o ed Resonant Circuit: LC Series Resonant Circuit<br>
A series resonant network consisting of a resistor of 30  $\Omega$ , a capacitor of 2  $\mu$ F, and an inductor<br>
sconnected across a sinusoidal supply voltage which has a constant o **ped Resonant Circuit:** LC Series Resonant Circuit<br>
te: A series resonant network consisting of a resistor of 30 G, a capacitor of 2  $\mu$ F, and an inductor<br>
is connected across a sinusoidal supply voltage which has a cons
	- a) Determine resonant frequency
	- b) Determine current at resonance
	- c) Determine voltage magnitude across inductor
	- d) Determine unloaded *Q* of circuit at resonance

#### **Solution**:

a) Resonant frequency:

$$
f_r = \frac{1}{2\pi\sqrt{LC}} = \frac{1}{2\pi\sqrt{0.002 \times 2 \times 10^{-6}}} = 2,516 \text{ Hz}
$$

![](_page_13_Figure_9.jpeg)

b) Current at resonant

$$
I = \frac{V}{R} = \frac{9}{30} = 0.3 \,\mathrm{A}
$$

c) Voltage across inductor

$$
|V_L| = I \times X_L = 0.3 \times (2\pi fL)
$$
  
= 0.3 \times (2\pi \times 2516 \times 0.002) = 9.48 V

![](_page_14_Figure_5.jpeg)

d) Determine unloaded *Q*:

$$
Q_u = \frac{\omega_0 L}{R} = \frac{1}{\omega_0 RC}
$$
  
=  $\frac{2\pi fL}{R} = \frac{2\pi \times 2516 \times 0.002}{30} = 1.05$  : not good Q-factor

![](_page_15_Picture_370.jpeg)

## **4 Review**

#### **Why the** *LC* **circuit (or** *LC***-tank) is called a tuned circuit or tank circuit?**

- The charge flows back and forth between the plates of the capacitor and through the inductor. The energy oscillates between a capacitor and an inductor until the internal resistance of the components and connecting wires makes the oscillations to die-out.
- The LC circuit behaves like a harmonic oscillator, akin to a pendulum swinging or water sloshing in a tank, which is why it's called a tuned or tank circuit.
- The LC circuit can act as an electrical resonator and storing energy oscillates between the electric field and magnetic field at the frequency called a resonant frequency.